

# LTM4600 DC/DC $\mu$ Module™ Thermal Performance

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## INTRODUCTION

The LTM4600 DC/DC  $\mu$ Module is a complete high power density stepdown regulator for 10A continuous (14A peak) loads. The  $\mu$ Module has two voltage options: 20V<sub>IN</sub> maximum for the LTM4600EV and 28V<sub>IN</sub> maximum for the LTM4600HVEV each housed in a small 15mm × 15mm × 2.8mm LGA surface mount package. Load current derating curves are provided in the datasheet for several input voltage, output voltage, and ambient temperatures with air flow. These derating curves provide guidelines for using the LTM4600 in ambient environments with regard to safe-operating-area (SOA). Also, there are efficiency curves in the datasheet that are used to extrapolate the power loss curves used in this thermal application note. The purpose of this thermal application note is to provide a guideline for using the  $\mu$ Module in ambient environments with or without air flow. The goal is to measure the temperature of a design, derive thermal models for different cases and finally determine the junction-to-ambient thermal resistance ( $\theta_{JA}$ ) in units of °C/W in the heat path. The data includes power loss curves, safe operating curves (SOA), thermal camera images and current derating curves verses ambient temperature with and without a heatsink. Also air flow will be included in the derating curves. The 24V designs are analyzed for a worse case temperature rise due to the lower efficiency exhibited in these higher input voltage designs.

## THERMAL MODEL

An example is shown in the schematic (Figure 1(a)), with a  $\mu$ Module attached to a 4-layer PCB with a size of 95mm × 76mm. To analyze this physical system, a simplified 1-D thermal model, which is presented in Figure 1(b), is employed to show the heat paths in the system. The heat is generated from the  $\mu$ Module and flows to the top and bottom sides. For the topside heat path,  $R_{JT}$  is used to represent the thermal resistance from junction to the top surface, while  $R_{TA}$  represents the resistance from the top surface to ambient. Similarly, for the bottom side,  $R_{JB}$  is the thermal resistance from junction to the bottom surface, and  $R_{BA}$  is the resistance from the bottom surface to ambient. The double-sided cooling scheme can be realized easily if heat sink is used for the top side.

## THERMAL IMAGING

### Case 1: No Heatsink

A 12V to 3.3V at 10A design and a 24V to 3.3V at 10A design are characterized for 33W operation at about 91% and 87% conversion efficiency respectively. This corresponds to a power loss of about 3W and 4.25W dissipated in the  $\mu$ Module and the PCB. The extra 4% loss on the 24V design is attributed to the extra power dissipation in the controller, and increased transition losses in the internal

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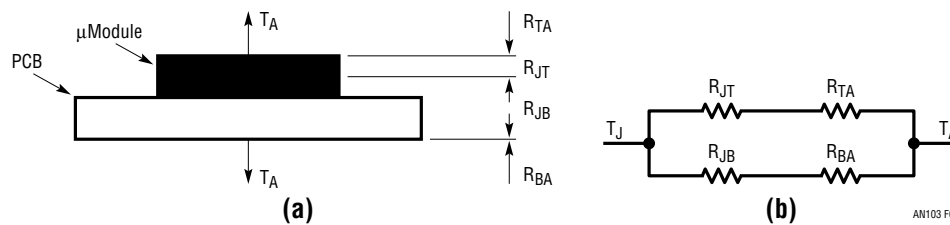


Figure 1. Thermal Model in The Design

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top MOSFET. This loss can be reduced by about 2%, or an efficiency of 89% from the 24V design, by connecting the EXT<sub>VCC</sub> pin to a 5V bias supply with a 50mA capability. The EXT<sub>VCC</sub> must be sequenced after the main input supply. Figure 2 shows a thermal image of the 12V to 3.3V design with several thermal image data points, and Figure 3 shows the 24V to 3.3V design with several thermal image data points. The maximum temperature in Figure 2 is equal to 66°C on the  $\mu$ Module with 3W of dissipation in the design, and Figure 3 has a maximum temperature of 82°C on the  $\mu$ Module with 4.25W of dissipation.

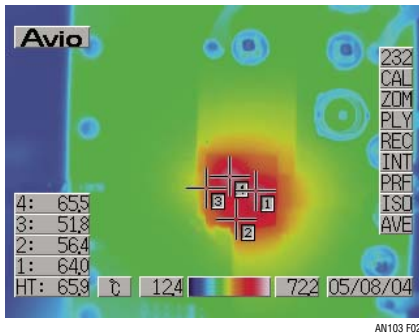
We only analyze a worse case with no heatsink and 4.25W of dissipation in Figure 3. Since it has a small top surface area on the  $\mu$ Module at 15mm  $\times$  15mm, the heat dissipation from the  $\mu$ Module topside can be ignored. So the thermal model shown in Figure 1 can be redrawn in Figure 4, which only has thermal resistances  $R_{JB}$  and  $R_{BA}$  at the

bottom side heat path. To measure the junction temperature of the  $\mu$ Module, a thermocouple is inserted into the  $\mu$ Module at a point close to the power MOSFET. This measured junction temperature is 89.8°C. The average temperature at the bottom side of the PCB is about 75°C. Therefore,  $R_{JB}$  and  $R_{BA}$  can be calculated at 3.5°C/W and 11.7°C/W respectively. The total thermal resistance from junction to ambient in this case is only 15.2°C/W.

## Case 2: With A BGA Heatsink

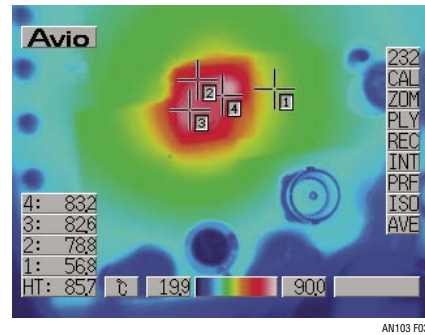
Figure 5 shows a thermal image with a surface mount BGA heatsink on top of the  $\mu$ Module. From the measurement, The  $T_{PCB}$  average temperature is about 54°C on the 12V to 3.3V design and about 73°C on the 24V to 3.3V design.

Figure 5 shows a side view of the LTM4600  $\mu$ Module with the surface mount BGA heatsink. Data point 2 indicates the heatsink temperature, and data point 4 indicates the joint



CONDITIONS: 25°C, NO AIR FLOW, NO HEATSINK, NO EXT<sub>VCC</sub>

Figure 2. LTM4600 12V to 3.3V at 10A, Top view



CONDITIONS: 25°C, NO AIR FLOW, NO HEATSINK, NO EXT<sub>VCC</sub>

Figure 3. LTM4600 24V to 3.3V at 10A, Top view

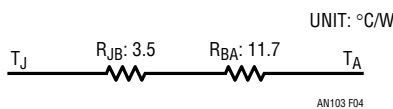
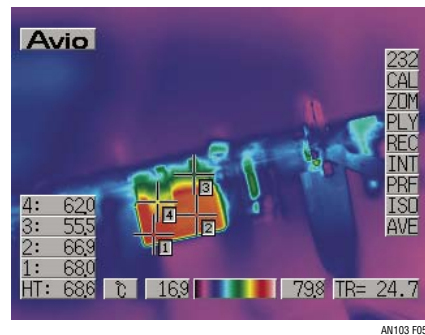


Figure 4. Thermal Model for Case 1 in Figure 3



CONDITIONS: 25°C, NO AIR FLOW, WAKEFIELD ENGINEERING PN# CIS20069, 15mm  $\times$  15mm  $\times$  9mm HEATSINK, NO EXT<sub>VCC</sub>

Figure 5. LTM4600 24V to 3.3V at 10A, Side View

point of the BGA heatsink and power  $\mu$ Module. The topside of the LTM4600  $\mu$ Module is very effective in transferring heat into an external heatsink. There is only a  $4^{\circ}\text{C}$  delta between the device and the heatsink with 4.25W of dissipation. The output current derating curves section will be discussed later with and without heatsinks under ambient conditions. The thermal model, which represents the case in Figure 5 with 4.25W of dissipation, is shown in Figure 6. In this case, the heat flows to both top and bottom sides. For topside heat path, the heat generated from the junction ( $R_{JH}$ ) to the  $\mu$ Module case, and then it reaches the heatsink and dissipates into ambient ( $R_{HA}$ ). For the bottom side heat path, the heat first flows to the 4-layer PCB before it dissipates to the ambient from the PCB. Here,  $R_{JB}$  is the thermal resistance from junction to PCB dissipation surface and it includes  $R_{JP}$  (junction to module pin) and  $R_{PB}$  (pin to PCB dissipation surface).

Since the heat sink temperature is about  $66^{\circ}\text{C}$  in Figure 5 and  $R_{HA}$  under natural convection condition can be ob-

tained at about  $21.5^{\circ}\text{C/W}$  from the datasheet of the manufacturer, we can know that the heat dissipation to topside is about 1.9W. The measured junction temperature in this case is about  $84^{\circ}\text{C}$ , so we can calculate all thermal resistances in the model as shown in Figure 6(b). Compared to the case without a heatsink in Figure 4, the heat spreading area to the bottom side in this case becomes smaller due to lower heat dissipation to bottom side, so the thermal resistances at bottom side heat path become larger in Figure 6. The total junction-to-ambient thermal resistance for this case with a BGA heatsink is about  $13.9^{\circ}\text{C/W}$ .

### Case 3: With A Metal Plate

Figure 7 shows the back side PCB view of a LTM4600 design that is mounted to a metal plate with a size of  $100\text{mm} \times 80\text{mm}$ . This thermal test case is analyzed for consideration of use in systems that desire back side PCB mounting of the power  $\mu$ Module. The power  $\mu$ Module can then be mounted to a metal carrier either directly or through a thermal conductive pad on a heatsink. This test

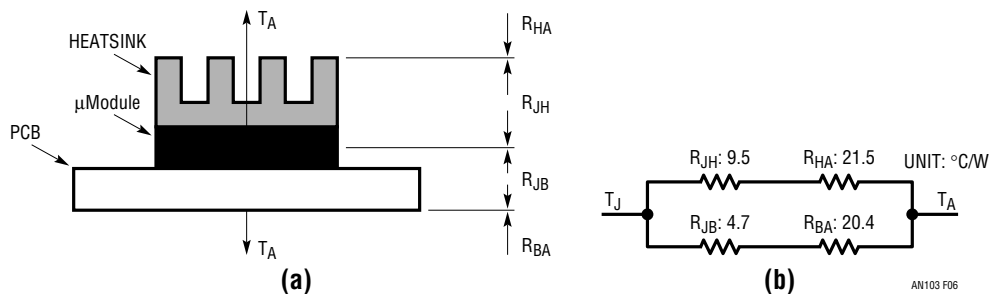
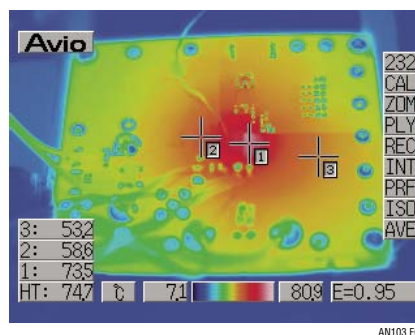


Figure 6. Thermal Model for Case 2



CONDITIONS: 24V TO 3.3V AT 10A,  $25^{\circ}\text{C}$ , NO AIR FLOW. A BERGQUIST "GAP PAD 1000" IS USED BETWEEN THE  $\mu$ Module AND THE METAL PLATE. 0.04 THICKNESS  $2^{\circ}\text{C/W}$ . (METAL PLATE =  $100\text{mm} \times 75\text{mm} \times 1.5\text{mm}$ )

Figure 7. LTM4600 24V to 3.3V at 10A, Back Side of the PCB

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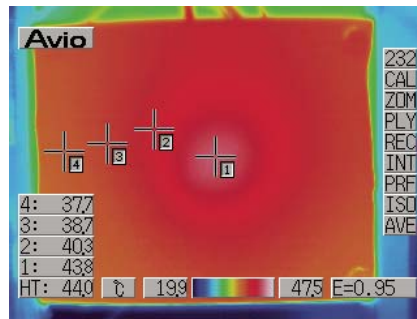
case uses a Bergquist “Gap Pad” for the thermal connection between the power  $\mu$ Module and metal carrier. The conditions are noted below Figure 7.

Figure 8 shows the metal plate view of the 33W design with the conditions noted below in the photo. The metal plate transfers heat effectively, and would provide an even better result under air flow. Similar to previous analysis, the average temperature of the bottom side of the PCB is about 66°C in Figure 7 and the average temperature of the metal plate is about 40°C in Figure 8. And the thermal resistance  $R_{MA}$  from metal plate to ambient is only about 7.5°C/W due to the large dissipation surface of the metal plate. The measured junction temperature is about 76°C. There is a thermal resistance drop from the top of the  $\mu$ Module to the metal plate. The Bergquist “Gap Pad” that is used between the  $\mu$ Module and the metal plate has a thermal resistance of 2°C/W, so this could be improved with the removal of the “Gap Pad”. The other 5°C/W thermal resistance drop is developed by the interface of the  $\mu$ Module and metal plate to the “Gap Pad”. This total

thermal resistance drop can be reduced by an improved thermal interface from the  $\mu$ Module to the metal plate. Here,  $R_{JM}$  is the total thermal resistance from junction to metal plate and it includes the thermal resistances from junction to dissipation surface of the metal plate:  $R_{JC}$  (junction to case),  $R_{PAD}$  (gap pad),  $R_{INTERFACE}$  (interfaces of case and metal plate to gap pad) and  $R_{METAL PLATE}$  (metal plate). We can obtain all thermal resistances as shown in Figure 9(b). In these thermal resistances, only  $R_{JC}$  (6°C/W to 9°C/W) and  $R_{JP}$  (1.5°C/W to 3°C/W) are dependent on the  $\mu$ Module and all other thermal resistances are related to specific customer designs. The total thermal resistance from junction to ambient in this case is about 12°C/W.

## DERATING CURVES VERSES AMBIENT TEMPERATURE AND AIR FLOW

Several derating curves are shown below to provide a guideline for the maximum load current that can be achieved at certain ambient temperatures. These curves



CONDITIONS: 24V TO 3.3V AT 10A, 25°C, NO AIR FLOW. A BERGQUIST “GAP PAD 1000” IS USED BETWEEN THE  $\mu$ Module AND THE METAL PLATE. 0.04 THICKNESS 2°C/W. (METAL PLATE = 100mm × 75mm × 1.5mm)

Figure 8. LTM4600 24V to 3.3V at 10A, Metal Plate View

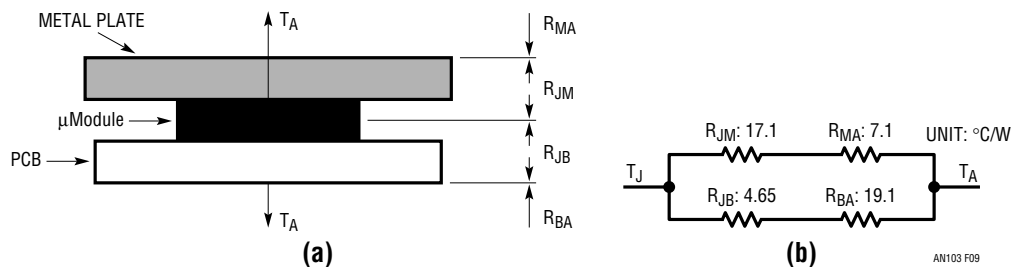


Figure 9. Thermal Model for Case 3

are characterized with 0LFM, 200LFM, and 400LFM. Also the curves are provided with heatsinks and no heatsinks. The power loss curves are provided to help establish an approximate  $\theta_{JA}$  for the characterized operating conditions that will ultimately be correlated to the thermal images above. The power loss curves and derating curves will be used to build a table to correlate our approximate  $\theta_{JA}$  and a reduced  $\theta_{JA}$  with increased air flow. We have chosen 5V, 12V, and 24V as the input operating conditions for this analysis. The two output voltages are 1.5V and 3.3V.

Figures 10 and 11 show the 1.5V and 3.3V power loss curves with load current and input voltages.

Figures 12, 13, and 14 are the three derating curves for 5V to 1.5V versus load current, air flow, and with and without heatsinks. Figures 15, 16, and 17 are the same derating curves for 12V to 1.5V. Figures 18, 19, and 20 are the derating curves for 24V to 1.5V. All of the curves are put into columns to designate the type of heatsink used in the test conditions.

Figures 21, 22 and 23 are the three derating curves for 12V to 3.3V at the different load currents, different air flow, and different heatsinks. Figures 24, 25, and 26 are the three derating curves for 24V to 3.3V. All of these curves are put into columns to designate the type of heatsink used in the test conditions.

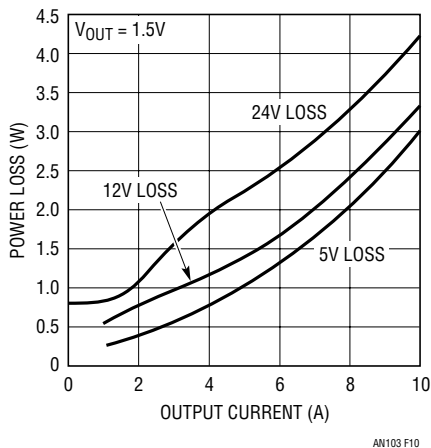


Figure 10. Power Loss vs Load Current

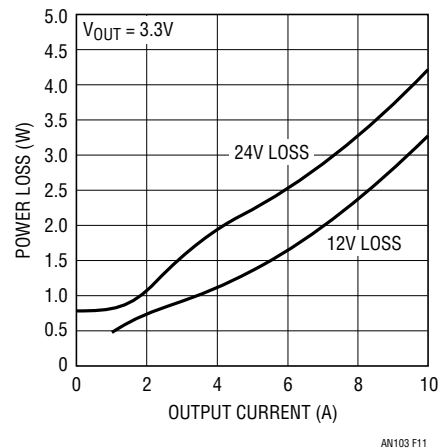


Figure 11. Power Loss vs Load Current

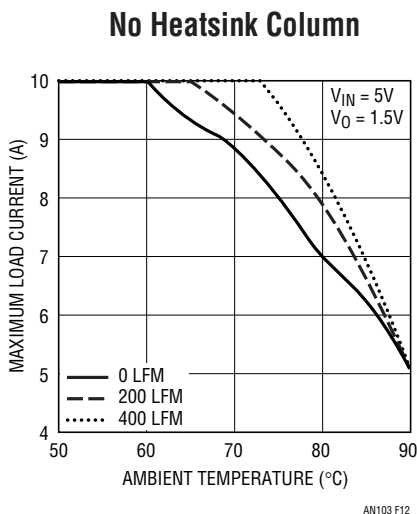


Figure 12

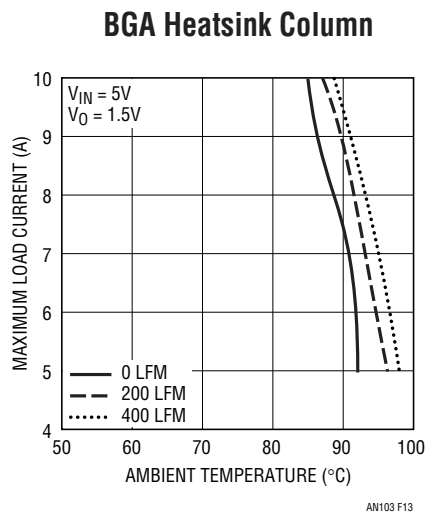


Figure 13

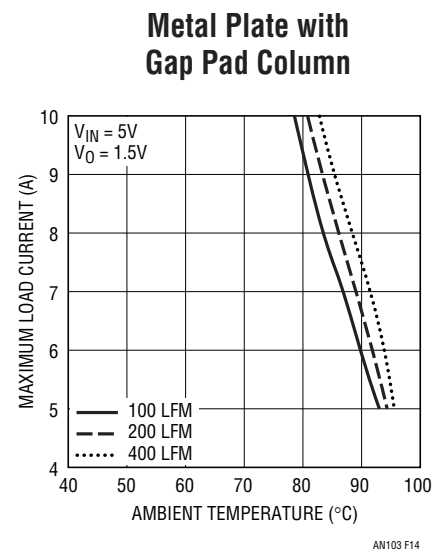
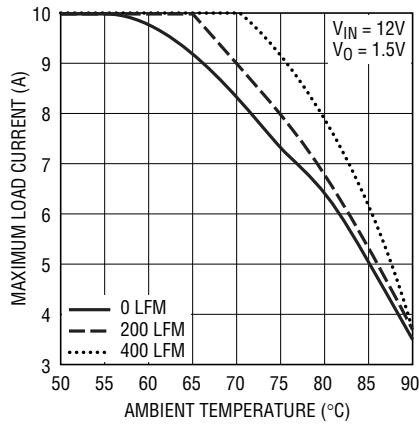


Figure 14

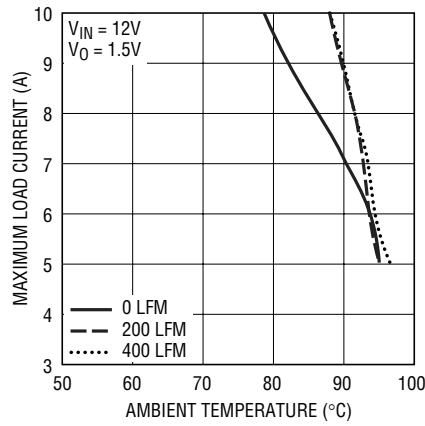
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**No Heatsink Column**



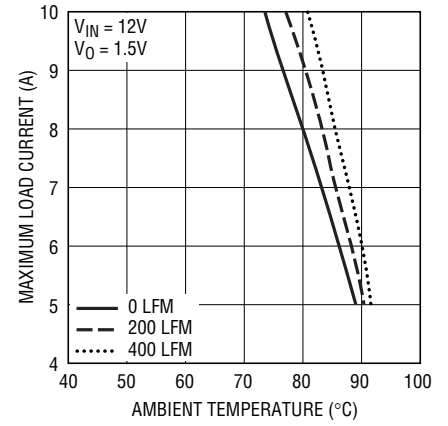
**Figure 15**

**BGA Heatsink Column**

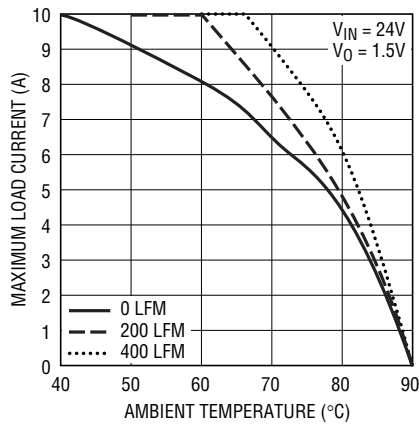


**Figure 16**

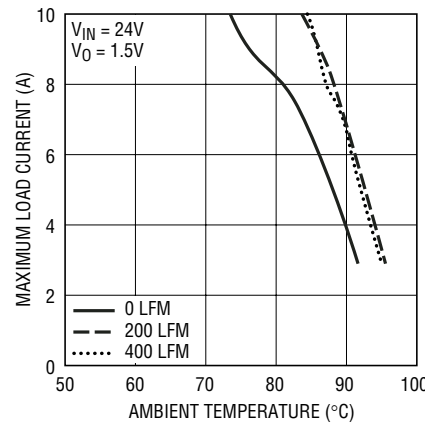
**Metal Plate with Gap Pad Column**



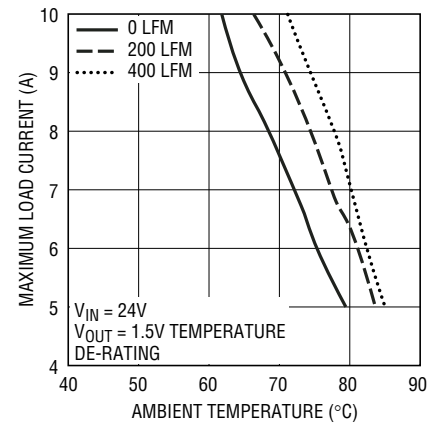
**Figure 17**



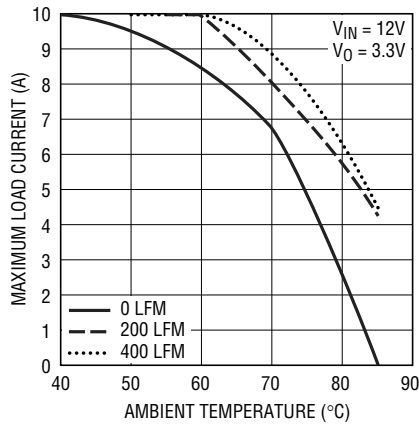
**Figure 18**



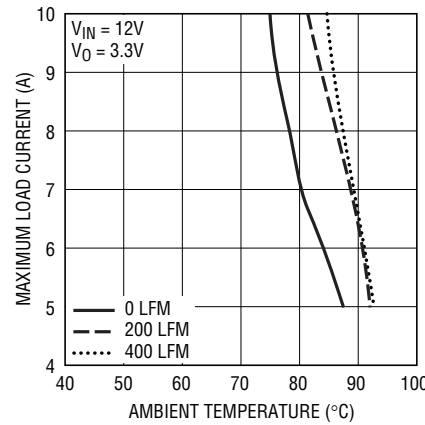
**Figure 19**



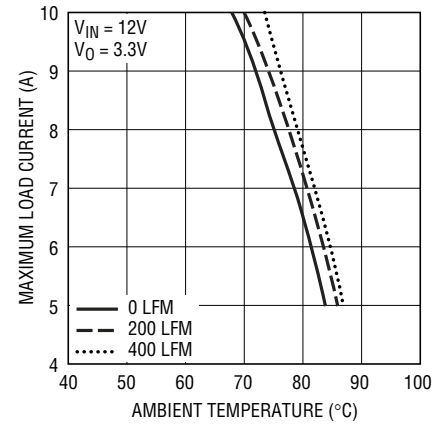
**Figure 20**



**Figure 21**



**Figure 22**



**Figure 23**

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### No Heatsink Column

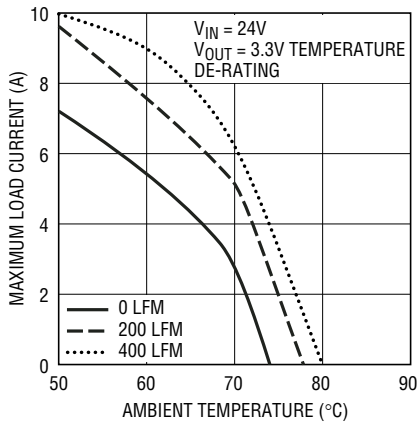


Figure 24

### BGA Heatsink Column

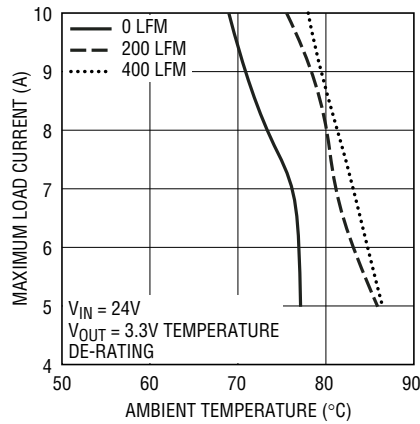


Figure 25

### Metal Plate with Gap Pad Column

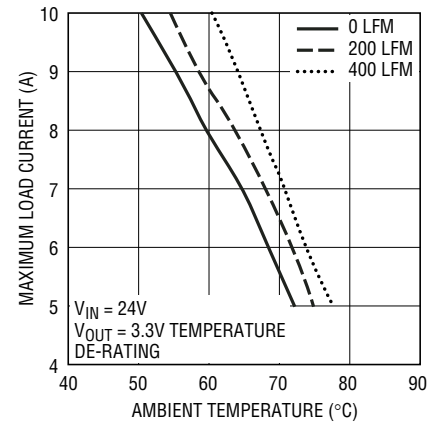


Figure 26

The power loss curves in Figures 10 and 11 will now be used in conjunction with the load current derating curves in Figures 12 through 26 to calculate an approximate  $\theta_{JA}$ . Each of the load current derating curves will lower the maximum load current as a function of the increased ambient temperature to keep the case temperature of the power  $\mu$ Module at 100°C maximum. This 100°C maximum is to allow for an increased rise of about 13°C to 20°C inside the  $\mu$ Module with a thermal resistance  $R_{JC}$  from junction to case at 6% to 9%°C/W. This will maintain the maximum operating temperature below 125°C. Each of the derating curves and the power loss curve that corre-

sponds to the correct output voltage can be used to solve for the approximate  $\theta_{JA}$  of the condition.

## CONCLUSION

The approximate  $\theta_{JA}$  was empirically solved for in the thermal image section of this application note. The data was taken with no air flow. The values for  $\theta_{JA}$  that were derived from the thermal model are 15.2°C/W, 13.9°C/W, and 12°C/W with no heatsink, a BGA heatsink, and a metal plate respectively. This data correlates very well with the zero air flow  $\theta_{JA}$  in Table 1 and Table 2.

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**Table 1. 1.5V Output**

DERATING CURVE	V <sub>IN</sub> (V)	POWER LOSS CURVE	AIR FLOW (LFM)	HEATSINK	θ <sub>JA</sub> (°C/W)
Figures 12, 15, 18	5, 12, 24	Figure 10	0	None	15.2
Figures 12, 15, 18	5, 12, 24	Figure 10	200	None	14
Figures 12, 15, 18	5, 12, 24	Figure 10	400	None	12
Figures 13, 16, 19	5, 12, 24	Figure 10	0	BGA Heatsink	13.9
Figures 13, 16, 19	5, 12, 24	Figure 10	200	BGA Heatsink	11.3
Figures 13, 16, 19	5, 12, 24	Figure 10	400	BGA Heatsink	10.25
Figures 14, 17, 20	5, 12, 24	Figure 10	0	Metal Plate	12
Figures 14, 17, 20	5, 12, 24	Figure 10	200	Metal Plate	9.5
Figures 14, 17, 20	5, 12, 24	Figure 10	400	Metal Plate	8.15

**Table 2. 3.3V Output**

DERATING CURVE	V <sub>IN</sub> (V)	POWER LOSS CURVE	AIR FLOW (LFM)	HEATSINK	θ <sub>JA</sub> (°C/W)
Figures 21, 24	12, 24	Figure 11	0	None	15.2
Figures 21, 24	12, 24	Figure 11	200	None	14.6
Figures 21, 24	12, 24	Figure 11	400	None	13.4
Figures 22, 25	12, 24	Figure 11	0	BGA Heatsink	13.9
Figures 22, 25	12, 24	Figure 11	200	BGA Heatsink	11.1
Figures 22, 25	12, 24	Figure 11	400	BGA Heatsink	10.5
Figures 23, 26	12, 24	Figure 11	0	Metal Plate	12
Figures 23, 26	12, 24	Figure 11	200	Metal Plate	10.8
Figures 23, 26	12, 24	Figure 11	400	Metal Plate	10.3

HEATSINK MANUFACTURER	PART NUMBER	PHONE NUMBER
Wakefield Engineering	CIS20069	603-635-2800
Bergquist Company	Gap Pad 1000SF	952-835-2322

A color version of this Application Note is available at [www.linear.com/micromodule](http://www.linear.com/micromodule)